## Sheet 2 - solution

1 A lossless transmission line is terminated with a $100 \Omega$ load. If the SWR on the line is 1.5, find the two possible values for the characteristic impedance of the line.

$$
\begin{aligned}
& |r|=\frac{S-1}{S+1}=\frac{0.5}{2.5}=0.2 \\
& |r|=\left|\frac{z_{L}-z_{0}}{z_{L}+z_{0}}\right|=\left|\frac{100-z_{0}}{100+z_{0}}\right| \quad\left(z_{0} \text { real }\right)
\end{aligned}
$$

So either,

$$
\frac{100-z_{0}}{100+z_{0}}=0.2 \Rightarrow z_{0}=z_{L} \frac{1-\Gamma}{1+\Gamma}=100\left(\frac{18}{1.2}\right)=66.7 \Omega
$$

$\sigma$

$$
\frac{100-z_{0}}{100+z_{0}}=-0.2 \Rightarrow z_{0}=z_{L} \frac{1-\Gamma}{1+\Gamma}=100\left(\frac{1.2}{.8}\right)=150 \Omega
$$

2 Let $Z_{s c}$ be the input impedance of a length of coaxial line when one end is shortcircuited and let $Z_{o c}$ be the input impedance of the line when one end is opencircuited. Derive an expression for the characteristic impedance of the cable in terms of $Z_{s c}$ and $Z_{o c}$.

$$
\begin{array}{ll}
z_{S C}=j z_{0} \tan \beta l & , z_{O C}=-j z_{0} \cot \beta l \\
z_{S C} \cdot z_{O C}=z_{0}^{2} \Rightarrow & z_{0}=\sqrt{z_{S C} z_{O C}}
\end{array}
$$

3 A $100 \Omega$ transmission line has an effective dielectric constant of 1.65. Find the shortest open-circuited length of this line that appears at its input as a capacitor of 5 pF at 2.5 GHz . Repeat for an inductance of 5 nH .

$$
\begin{gathered}
C: \quad Z_{o c}=-j / \omega c=-j 12.73 \Omega=-j z_{0} \cot \beta l \quad c=5 \rho F \\
\tan \beta l=100 / 12.73 \Rightarrow \beta l=82.74^{\circ} \mathrm{V} \\
\lambda_{0}=0.12 \mathrm{~m}, \beta=2 \pi \sqrt{\epsilon e} / \lambda_{0}=38540 / \mathrm{m} \Rightarrow l=2.147 \mathrm{~cm} \\
L: \quad Z_{0 c}=j \omega L=+j 78.5 \Omega=-j z_{0} \cot \beta l \quad L=5 \mathrm{nH} \\
\tan \beta l=-100 / 78.5 \Rightarrow \beta l=128.1^{\circ} \mathrm{V} \Rightarrow l=3.324 \mathrm{~cm}
\end{gathered}
$$

These results were verified with Serenade.

4 A radio transmitter is connected to an antenna having an impedance $80+\mathrm{j} 40 \Omega$ with a $50 \Omega$ coaxial cable. If the $50 \Omega$ transmitter can deliver 30 W when connected to a $50 \Omega$ load, how much power is delivered to the antenna?

$$
\begin{aligned}
& \Gamma=\frac{z_{L-}-z_{0}}{z_{L}+z_{0}}=\frac{30+j 40}{130+j 40}=\frac{50153^{\circ}}{136 / 17^{\circ}}=0.367 \angle 36^{\circ} \\
& P_{L O A D}=P_{I N C}-P_{R E F}=P_{I N C}\left(1-|\Gamma|^{2}\right)=30\left[1-(.367)^{2}\right]=25.9 \mathrm{~W}
\end{aligned}
$$

5 The transmission line circuit shown below has $\mathrm{V}_{\mathrm{g}}=15 \mathrm{Vrms}, \mathrm{Z}_{\mathrm{g}}=75 \Omega, \mathrm{Z}_{\mathrm{o}}=75 \Omega, \mathrm{Z}_{\mathrm{L}}=$ $60-\mathrm{j} 40 \Omega$, and $\ell=0.7 \lambda$. Compute the power delivered to the load using three different techniques:
(a) find $\Gamma$ and compute

$$
P_{L}=\left(\frac{V_{g}}{2}\right)^{2} \frac{1}{Z_{0}}\left(1-|\Gamma|^{2}\right) ;
$$

$$
V_{g}=15 \mathrm{v} \mathrm{Rm} \mathrm{~s}, \quad Z_{g}=75 \Omega, \quad Z_{0}=75 \Omega, \quad z_{L}=60-j 40 \Omega, l=0.7 \lambda .
$$

a)

$$
\begin{aligned}
& \Gamma=\frac{z_{1}-z_{0}}{z_{L}+z_{0}}=\frac{-15-j 40}{135-j 40}=\frac{42.7 /-10.5^{\circ}}{140.8 / 16.5^{\circ}}=0.303 /-94^{\circ}=-0.021-j 0.302 \\
& P_{L}=\left(\frac{V_{g}}{2}\right)^{2} \frac{1}{z_{0}}\left(1-1 r 1^{2}\right)=0.681 \mathrm{~W}
\end{aligned}
$$

This method is actually based on $P_{L}=P_{\text {inc }}\left(1-|\Gamma|^{2}\right), \alpha+i s$ the simplest method, but only applies to locales lines.
(b) find $Z_{\text {in }}$ and compute

$$
P_{L}=\left|\frac{V_{g}}{Z_{g}+Z_{\text {in }}}\right|^{2} \operatorname{Re}\left(Z_{\text {in }}\right) ; \text { and }
$$

b) $z_{\text {in }}=z_{0} \frac{z_{L}+j z_{0} \tan \beta l}{z_{0}+j z_{L} \tan \beta l}=75 \frac{60+j 190.8}{198.1+j 184.7}=75 \frac{200 / 72.5^{\circ}}{270.8143^{\circ}}$

$$
=55.4 \angle 29.5^{\circ}=48.2+j 27.3 \Omega
$$

$$
P_{L}=\left|\frac{V_{g}}{z_{g}+z_{i n}}\right|^{2} R_{e}\left(z_{i n}\right)=\left|\frac{15}{123.2+j 27.3}\right|^{2}(48.2)=0.681 \mathrm{wr}
$$

This method computes $P_{L}=P_{\text {in }}=\left|I_{i n}\right|^{2} R_{i n}$, and also agghlies only to looses lines.
(c) find $V_{L}$ and compute

$$
P_{L}=\left|\frac{V_{L}}{Z_{L}}\right|^{2} \operatorname{Re}\left(Z_{L}\right) .
$$

$$
\begin{aligned}
V(z) & =V^{+}\left(e^{-j \beta z}+\Gamma e^{j \beta z}\right) \\
V_{L} & =V(0)=V^{+}(1+\Gamma) \quad V^{+}=\frac{V_{g}}{2}=7.5 v \\
& =7.5(1-.021-j .302) \\
& =7.68 \frac{1-17^{0}}{} \\
P_{L} & =\left|\frac{V_{L}}{z_{L}}\right|^{2} \operatorname{Re}\left(z_{L}\right)=\left(\frac{7.68}{72.1}\right)^{2}(60)=0.681 \mathrm{w}
\end{aligned}
$$

(d) Discuss the rationale for each of these methods. Which of these methods can be used if the line is not lossless?


This method computes $P_{L}=\left|I_{L}\right|^{2} R_{L}$, and applies to lossy as well as lossless hines. Note the concept that $v^{+}=V_{g} / 2$ requires a good understanding of the transmission line equations, and only applies here because $z_{g}=z_{0}$.

6 For a purely reactive load impedance of the form $\mathrm{Z}_{\mathrm{L}}=\mathrm{jX}$, show that the reflection coefficient magnitude If $I \Gamma I$ is always unity. Assume the characteristic impedance $Z_{0}$ is real.

$$
\begin{aligned}
& z_{L}=j x \\
& \Gamma=\frac{z_{L}-z_{0}}{z_{L}+z_{0}}=\frac{j x-z_{0}}{j x+z_{0}} \\
& |\Gamma|^{2}=\Gamma \Gamma^{*}=\frac{\left(j x-z_{0}\right)}{\left(j x+z_{0}\right)} \frac{\left(-j x-z_{0}\right)}{\left(-j x+z_{0}\right)}=\frac{x^{2}-j z_{0} x+j z_{0} x+z_{0}^{2}}{x^{2}+z_{0}^{2}}=1 \mathrm{~V}
\end{aligned}
$$

7 Consider the transmission line circuit shown below. Compute the incident power, the reflected power, and the power transmitted into the infinite $75 \Omega$ line. Show that power conservation is satisfied.


POWER DELVERED BY SOURCE $=\frac{1}{2} \frac{(10)^{2}}{50+75}=0.400 \mathrm{~W}$
Power DISSIPATED in 50 LO LOAD $=\frac{1}{2}(50)\left(\frac{10}{50+75}\right)^{2}=0.160 \mathrm{w}$
POWER TRANSMITTED DOWN LINE $=\frac{1}{2}(75)\left(\frac{10}{50+75}\right)^{2}=0.240 \mathrm{~W}$
INCIDENT POWER $=\frac{1}{2}(50)\left(\frac{10}{50+50}\right)^{2}=0.250 \mathrm{~W} \quad \checkmark$
REFLECTED POWER $=P_{\text {INC }}|\Gamma|^{2}=.250\left|\frac{75-50}{75+50}\right|^{2}=0.010 \mathrm{~W} \quad \checkmark$
— $P_{I N C}-P_{\text {REF }}=.250-.010=0.240=P_{\text {TRANS }}$
$P_{\text {DISS }}+P_{\text {TRANS }}=.160+.240=0.400=P_{\text {SOURCE }}$

8 A generator is connected to a transmission line as shown below. Find the voltage as a function of $z$ along the transmission line. Plot the magnitude of this voltage for
$-\ell \leq \mathrm{Z} \leq 0$


$$
\begin{aligned}
& \Gamma=\frac{-20-j 40}{180-j 40}=\frac{44.7 L-116.6^{\circ}}{184.4 L-12.5^{\circ}}=0.24 L-104^{\circ} \\
& V_{L}=10 \frac{80-j 40}{180-j 40}=10 \frac{89.4 L-26^{\circ}}{184 L^{\circ}}=4.058-j 0.233 \\
& V(z)=V^{+}\left[e^{-j \beta z}+\Gamma e^{j \beta}\right] \quad V^{+}=10 \frac{100}{100+100}=5 v
\end{aligned}
$$

So

$$
\begin{aligned}
& V(z)=5\left[e^{-j z}+\Gamma p j \beta\right] \\
& V_{\text {MAX }}=5(1+|\Gamma|)=5(1.24)=6.2 \text { at } z=-0.355 \lambda \\
& V_{\text {MIN }}=5(|-|\Gamma|)=5(.76)=3.8 \text { at } z=-0.105 \lambda
\end{aligned}
$$

These results repeat evens $\lambda / 2$.
$|V(z)|$ is plotted below


Good Luck
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